

# Features

- ➤ Fast charge and conditioning of nickel cadmium or nickel-metal hydride batteries
- ➤ Hysteretic PWM switch-mode current regulation or gated control of an external regulator
- ➤ Easily integrated into systems or used as a stand-alone charger
- ➤ Pre-charge qualification of temperature and voltage
- Configurable, direct LED outputs display battery and charge status
- Fast-charge termination by Δ temperature/Δ time, peak volume detection, -ΔV, maximum voltage, maximum temperature, and maximum time
- Optional top-off charge and pulsed current maintenance charging
- ➤ Logic-level controlled low-power mode (< 5µA standby current)

## **General Description**

The bq2004E and bq2004H Fast Charge ICs provide comprehensive fast charge control functions together with high-speed switching power control circuitry on a monolithic CMOS device

Integration of closed-loop current control circuitry allows the bq2004 to be the basis of a cost-effective solution for stand-alone and systemintegrated chargers for batteries of one or more cells.

Switch-activated discharge-beforecharge allows bq2004E/H-based chargers to support battery conditioning and capacity determination.

High-efficiency power conversion is accomplished using the bq2004E/H as a hysteretic PWM controller for switch-mode regulation of the charging current. The bq2004E/H may alternatively be used to gate an externally regulated charging current.

Fast charge may begin on application of the charging supply, replacement of the battery, or switch depression. For safety, fast charge is inhibited unless/until the battery temperature and voltage are within configured limits.

**Fast-Charge ICs** 

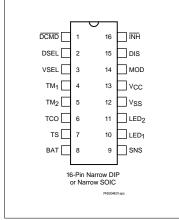
Temperature, voltage, and time are monitored throughout fast charge. Fast charge is terminated by any of the following:

- Rate of temperature rise (ΔT/Δt)
- Peak voltage detection (PVD)
- Negative delta voltage  $(-\Delta V)$
- Maximum voltage
- Maximum temperature
- Maximum time

After fast charge, optional top-off and pulsed current maintenance phases with appropriate display mode selections are available.

The bq2004H differs from the bq2004E only in that fast charge, hold-off, and top-off time units have been scaled up by a factor of two, and the bq2004H provides different display selections. Timing differences between the two ICs are illustrated in Table 1. Display differences are shown in Table 2.

## **Pin Connections**



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### **Pin Names**

$\overline{\mathrm{DCMD}}$	Discharge command	SNS	Sense resistor input
DSEL	Display select	$\mathrm{LED}_1$	Charge status output 1
VSEL	Voltage termination	$\mathrm{LED}_2$	Charge status output 2
	select	$V_{\rm SS}$	System ground
$\mathrm{TM}_1$	Timer mode select 1	$V_{\rm CC}$	$5.0 \text{V} \pm 10\%$ power
$TM_2$	Timer mode select 2	MOD	Charge current control
TCO	Temperature cutoff	DIS	Discharge control
TS	Temperature sense		output
BAT	Battery voltage	$\overline{\text{INH}}$	Charge inhibit input

## **Pin Descriptions**

### DCMD Discharge-before-charge control input

The  $\overline{DCMD}$  input controls the conditions that enable discharge-before-charge.  $\overline{DCMD}$  is pulled up internally. A negative-going pulse on  $\overline{DCMD}$  initiates a discharge to endof-discharge voltage (EDV) on the BAT pin,  $\overline{DCMD}$  to ground enables automatic discharge-before-charge on every new charge cycle start.

#### DSEL Display select input

This three-state input configures the charge status display mode of the  $\rm LED_1$  and  $\rm LED_2$  outputs and can be used to disable top-off and pulsed-trickle. See Table 2.

#### VSEL Voltage termination select input

This three-state input controls the voltage-termination technique used by the bq2004E/H. When high, PVD is active. When floating,  $\Delta V$  is used. When pulled low, both PVD and  $\Delta V$  are disabled.

# $\begin{array}{ll} TM_1- & Timer\ mode\ inputs \\ TM_2 & \end{array}$

 $TM_1$  and  $TM_2$  are three-state inputs that configure the fast charge safety timer, voltage termination hold-off time, "top-off", and trickle charge control. See Table 1.

### TCO Temperature cut-off threshold input

Input to set maximum allowable battery temperature. If the potential between TS and SNS is less than the voltage at the TCO input, then fast charge or top-off charge is terminated.

### TS Temperature sense input

Input, referenced to SNS, for an external thermister monitoring battery temperature.

#### BAT Battery voltage input

BAT is the battery voltage sense input, referenced to SNS. This is created by a high-impedance resistor-divider network connected between the positive and the negative terminals of the battery.

#### SNS Charging current sense input

SNS controls the switching of MOD based on an external sense resistor in the current path of the battery. SNS is the reference potential for both the TS and BAT pins. If SNS is connected to Vss, then MOD switches high at the beginning of charge and low at the end of charge.

# LED<sub>1</sub>- Charge status outputs LED<sub>2</sub>

Push-pull outputs indicating charging status. See Table 2.

#### Vss Ground

### $V_{CC} \hspace{0.5cm} V_{CC} \hspace{0.5cm} \text{supply input} \\$

 $5.0V, \pm 10\%$  power input.

#### MOD Charge current control output

MOD is a push-pull output that is used to control the charging current to the battery. MOD switches high to enable charging current to flow and low to inhibit charging current flow

### DIS Discharge control output

Push-pull output used to control an external transistor to discharge the battery before charging.

#### **INH** Charge inhibit input

When low, the bq2004E/H suspends all charge actions, drives all outputs to high impedance, and assumes a low-power operational state. When transitioning from low to high, a new charge cycle is started.

## **Functional Description**

Figure 2 shows a block diagram and Figure 3 shows a state diagram of the bq2004E/H.

# **Battery Voltage and Temperature Measurements**

Battery voltage and temperature are monitored for maximum allowable values. The voltage presented on the battery sense input, BAT, should represent a two-cell potential for the battery under charge. A resistor-divider ratio of:

$$\frac{RB1}{RB2} = \frac{N}{2} - 1$$

is recommended to maintain the battery voltage within the valid range, where N is the number of cells, RB1 is the resistor connected to the positive battery terminal, and RB2 is the resistor connected to the negative battery terminal. See Figure 1.

Note: This resistor-divider network input impedance to end-to-end should be at least  $200k\Omega$  and less than  $1M\Omega$ .

A ground-referenced negative temperature coefficient thermistor placed in proximity to the battery may be used as a low-cost temperature-to-voltage transducer. The temperature sense voltage input at TS is developed using a resistor-thermistor network between  $V_{\rm CC}$  and  $V_{\rm SS}$ . See Figure 1. Both the BAT and TS inputs are referenced to SNS, so the signals used inside the IC are:

$$V_{\rm BAT}$$
 -  $V_{\rm SNS}$  =  $V_{\rm CELL}$  
$$and$$
 
$$V_{\rm TS}$$
 -  $V_{\rm SNS}$  =  $V_{\rm TEMP}$ 

## **Discharge-Before-Charge**

The  $\overline{DCMD}$  input is used to command discharge-before-charge via the DIS output. Once activated, DIS becomes active (high) until  $V_{CELL}$  falls below  $V_{EDV}$ , at which time DIS goes low and a new fast charge cycle begins.

The  $\overline{DCMD}$  input is internally pulled up to  $V_{CC}$  (its inactive state). Leaving the input unconnected, therefore, results in disabling discharge-before-charge. A negative going pulse on  $\overline{DCMD}$  initiates discharge-before-charge at any time regardless of the current state of the bq2004. If  $\overline{DCMD}$  is tied to  $V_{SS}$ , discharge-before-charge will be the first step in all newly started charge cycles.

## **Starting A Charge Cycle**

A new charge cycle is started by:

- 1. Application of  $V_{CC}$  power.
- 2.  $V_{CELL}$  falling through the maximum cell voltage,  $V_{MCV}$  where:

$$V_{MCV} = 0.8*V_{CC} \pm 30mV$$

3. A transition on the  $\overline{\text{INH}}$  input from low to high.

If  $\overline{\text{DCMD}}$  is tied low, a discharge-before-charge will be executed as the first step of the new charge cycle. Otherwise, pre-charge qualification testing will be the first step.

The battery must be within the configured temperature and voltage limits before fast charging begins.

The valid battery voltage range is  $V_{\rm EDV}$  <  $V_{\rm BAT}$  <  $V_{\rm MCV}$  where:

$$V_{EDV} = 0.4 * V_{CC} \pm 30 mV$$

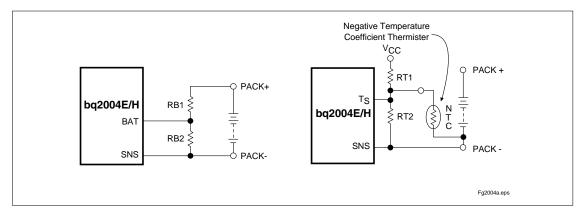


Figure 1. Voltage and Temperature Monitoring

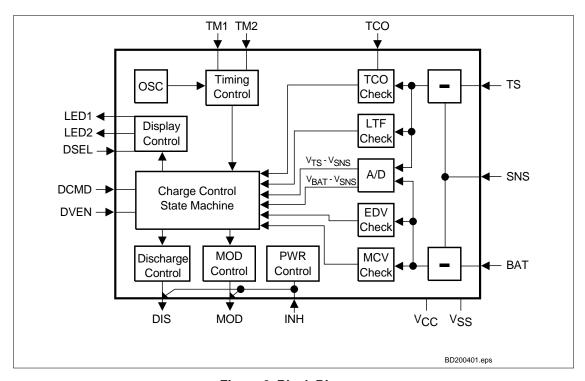


Figure 2. Block Diagram

The valid temperature range is  $V_{\rm HTF} < V_{\rm TEMP} < V_{\rm LTF}$ , where:

$$V_{LTF} = 0.4 * V_{CC} \pm 30 mV$$
 
$$V_{HTF} = [(1/3 * V_{LTF}) + (2/3 * V_{TCO})] \pm 30 mV$$

 $V_{TCO}$  is the voltage presented at the TCO input pin, and is configured by the user with a resistor divider between  $V_{CC}$  and ground. The allowed range is 0.2 to 0.4  $\ast\,V_{CC}$ .

If the temperature of the battery is out of range, or the voltage is too low, the chip enters the charge pending state and waits for both conditions to fall within their allowed limits. During the charge-pending mode, the IC first applies a top-off charge to the battery.

The top-off charge, at the rate of 1/8 of the fast charge, continues until the fast-charge conditions are met or the top-off time-out period is exceeded. The IC then trickle charges until the fast-charge conditions are met. There is no time limit on the charge pending state; the charger remains in this state as long as the voltage or temperature conditions are outside of the allowed limits. If the voltage is too high, the chip goes to the battery absent state and waits until a new charge cycle is started.

Fast charge continues until termination by one or more of the six possible termination conditions:

- Delta temperature/delta time  $(\Delta T/\Delta t)$
- Peak voltage detection (PVD)
- Negative delta voltage (-∆V)
- Maximum voltage
- Maximum temperature
- Maximum time

### PVD and - AV Termination

The bq2004E/H samples the voltage at the BAT pin once every 34s. When - $\Delta V$  termination is selected, if  $V_{\rm CELL}$  is lower than any previously measured value by 12mV  $\pm 4 {\rm mV}$  (6mV/cell), fast charge is terminated. When PVD termination is selected, if  $V_{\rm CELL}$  is lower than any previously measured value by 6mV  $\pm 2 {\rm mV}$  (3mV/cell), fast charge is terminated. The PVD and - $\Delta V$  tests are valid in the range 0.4 \*  $V_{\rm CC}$  <  $V_{\rm CELL}$  < 0.8 \*  $V_{\rm CC}$ 

VSEL Input	Voltage Termination		
Low	Disabled		
Float	$-\Delta V$		
High	PVD		

## **Voltage Sampling**

Each sample is an average of voltage measurements. The IC takes 32 measurements in PVD mode and 16 measurements in - $\!\Delta V$  mode. The resulting sample periods (9.17ms and 18.18ms, respectively) filter out harmonics centered around 55Hz and 109Hz. This technique minimizes the effect of any AC line ripple that may feed through the power supply from either 50Hz or 60Hz AC sources. Tolerance on all timing is  $\pm 16\%$ .

# Temperature and Voltage Termination Hold-off

A hold-off period occurs at the start of fast charging. During the hold-off period,  $\text{-}\Delta V$  and  $\Delta T/\Delta t$  termination are disabled. The MOD pin is enabled at a duty cycle of 260µs active for every 1820µs inactive. This modulation results in an average rate 1/8th that of the fast charge rate. This avoids premature termination on the voltage spikes sometimes produced by older batteries when fast-charge current is first applied. Maximum voltage and maximum temperature terminations are not affected by the hold-off period.

#### **∆T/∆t Termination**

The bq2004E/H samples at the voltage at the TS pin every 34s, and compares it to the value measured two samples earlier. If  $V_{TEMP}$  has fallen  $16mV~\pm 4mV$  or more, fast charge is terminated. The  $\Delta T/\Delta t$  termination test is valid only when  $V_{TCO} < V_{TEMP} < V_{LTF}.$ 

### **Temperature Sampling**

Each sample is an average of 16 voltage measurements. The resulting sample period (18.18ms) filters out harmonics around 55Hz. This technique minimizes the effect of any AC line ripple that may feed through the power supply from either 50Hz or 60Hz AC sources. Tolerance on all timing is  $\pm 16\%$ .

### Maximum Voltage, Temperature, and Time

Anytime  $V_{CELL}$  rises above  $V_{MCV}$ , the LEDs go off and current flow into the battery ceases immediately. If  $V_{CELL}$  then falls back below  $V_{MCV}$  before  $t_{MCV}=1.5s~\pm0.5s$ , the chip transitions to the Charge Complete state (maximum voltage termination). If  $V_{CELL}$  remains above  $V_{MCV}$  at the expiration of  $t_{MCV}$ , the bq2004E/H transitions to the Battery Absent state (battery removal). See Figure 3.

Maximum temperature termination occurs anytime  $V_{TEMP}$  falls below the temperature cutoff threshold  $V_{TCO}$ . Charge will also be terminated if  $V_{TEMP}$  rises above the low temperature fault threshold,  $V_{LTF}$ , after fast charge begins.

Table 1. Fast Charge Safety Time/Hold-Off/Top-Off Table

Corresp Fast-C	harge			Fast-C Sat	oical Charge fety (min)	PVD	Í-Off	Top-Off Rate		•		Pulse- Trickle		se- ckle d (Hz)
2004E	2004H	TM1	TM2	2004E	2004H	2004E	2004H	2004E	2004H	Rate	2004E	2004H		
C/4	C/8	Low	Low	325	650	137	273	Disa	abled	Disabled	Disa	bled		
C/2	C/4	Float	Low	154	325	546	546	Disa	abled	C/512	15	30		
1C	C/2	High	Low	77	154	273	546	Disa	abled	C/512	7.5	15		
2C	1C	Low	Float	39	77	137	273	Disa	abled	C/512	3.75	7.5		
4C	2C	Float	Float	19	39	68	137	Disa	abled	C/512	1.88	3.75		
C/2	C/4	High	Float	154	325	546	546	C/16	C/32	C/512	15	30		
1C	C/2	Low	High	77	154	273	546	C/8	C/16	C/512	7.5	15		
2C	1C	Float	High	39	77	137	273	C/4	C/18	C/512	3.75	7.5		
4C	2C	High	High	19	39	68	137	C/2	C/4	C/512	1.88	3.75		

**Note:** Typical conditions = 25°C,  $V_{CC} = 5.0V$ .

Table 2. bq2004E/H LED Output Summary

Mode 1 bq2004E	Charge Action State	LED <sub>1</sub>	LED <sub>2</sub>
	Battery absent	Low	Low
DSEL = V <sub>SS</sub>	Fast charge pending or a discharge-before-charge in progress	High	High
DSEL = VSS	Fast charging	Low	High
	Fast charge complete, top-off, and/or trickle	High	Low
Mode 1 bq2004H	Charge Action State	LED <sub>1</sub>	LED <sub>2</sub>
	Battery absent	Low	Low
	Discharge-before-charge in progress	High	High
$DSEL = V_{SS}$	Fast charge pending	Low	1/8 second high 1/8 second low
	Fast charging	Low	High
	Fast charge complete, top-off, and/or trickle	High	Low
Mode 2 bq2004E	Charge Action State (See note)	LED <sub>1</sub>	LED <sub>2</sub>
	Battery absent	Low	Low
DSEL = Floating	Fast charge pending or discharge-before-charge in progress	High	High
DSEL = Floating	Fast charging	Low	High
	Fast charge complete, top-off, and/or trickle	High	Low
Mode 2 bq2004H	Charge Action State (See note)	LED <sub>1</sub>	LED <sub>2</sub>
	Battery absent	Low	Low
	Discharge-before-charge in progress	High	High
DSEL = Floating	Fast charge pending	Low	1/8 second high 1/8 second low
	Fast charging	Low	High
	Fast charge complete, top-off, and/or trickle	High	Low
Mode 3 bq2004E/H	Charge Action State	LED <sub>1</sub>	LED <sub>2</sub>
	Battery absent	Low	Low
$DSEL = V_{CC}$	Fast charge pending or discharge-before-charge in progress	Low	1/8 second high 1/8 second low
	Fast charging	Low	High
	Fast charge complete, top-off, and/or trickle	High	Low

**Note:** Pulse trickle is inhibited in Mode 2.

Maximum charge time is configured using the TM pin. Time settings are available for corresponding charge rates of C/4, C/2, 1C, and 2C. Maximum time-out termination is enforced on the fast-charge phase, then reset, and enforced again on the top-off phase, if selected. There is no time limit on the trickle-charge phase.

#### **Top-off Charge**

An optional top-off charge phase may be selected to follow fast charge termination for the C/2 through 4C rates. This phase may be necessary on NiMH or other battery chemistries that have a tendency to terminate charge prior to reaching full capacity. With top-off enabled, charging continues at a reduced rate after fast-charge termination for a period of time equal to 0.235\* the fast-charge safety time (See Table 1.) During top-off, the MOD pin is enabled at a duty cycle of 260µs active for every 1820µs inactive. This modulation results in an average rate 1/8th that of the fast charge rate. Maximum voltage, time, and temperature are the only termination methods enabled during top-off.

### **Pulse-Trickle Charge**

Pulse-trickle charging may be configured to follow the fast charge and optional top-off charge phases to compensate for self-discharge of the battery while it is idle in the charger.

In the pulse-trickle mode, MOD is active for 260 $\mu$ s of a period specified by the settings of TM1 and TM2. See Table 1. The resulting trickle-charge rate is C/512. Both pulse trickle and top-off may be disabled by tying TM1 and TM2 to V<sub>SS</sub> or by selecting Mode 2 in the display.

## **Charge Status Indication**

Charge status is indicated by the LED<sub>1</sub> and LED<sub>2</sub> outputs. The state of these outputs in the various charge cycle phases is given in Table 2 and illustrated in Figure 3.

In all cases, if  $V_{\rm CELL}$  exceeds the voltage at the MCV pin, both LED<sub>1</sub> and LED<sub>2</sub> outputs are held low regardless of other conditions. Both can be used to directly drive an LED.

### **Charge Current Control**

The bq2004E/H controls charge current through the MOD output pin. The current control circuitry is designed to support implementation of a constant-current switching regulator or to gate an externally regulated current source.

When used in switch mode configuration, the nominal regulated current is:

$$I_{REG} = 0.225 V/R_{SNS}$$

Charge current is monitored at the SNS input by the voltage drop across a sense resistor, R<sub>SNS</sub>, between the low side of the battery pack and ground. R<sub>SNS</sub> is sized to provide the desired fast charge current.

If the voltage at the SNS pin is less than  $V_{\rm SNSLO}$ , the MOD output is switched high to pass charge current to the battery.

When the SNS voltage is greater than  $V_{\rm SNSHI}$ , the MOD output is switched low—shutting off charging current to the battery.

$$\begin{split} V_{SNSLO} &= 0.04 * V_{CC} \pm 25 mV \\ V_{SNSHI} &= 0.05 * V_{CC} \pm 25 mV \end{split}$$

When used to gate an externally regulated current source, the SNS pin is connected to  $V_{\rm SS}$ , and no sense resisitor is required.

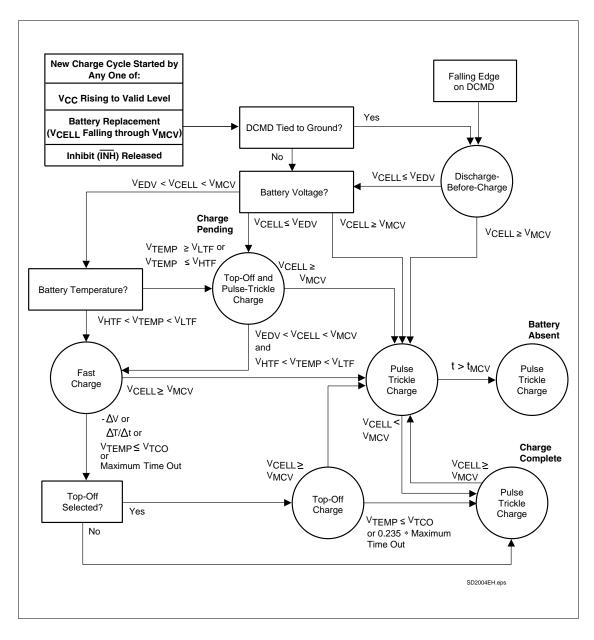


Figure 3. Charge Algorithm State Diagram

# **Absolute Maximum Ratings**

Symbol	Parameter	Minimum	Maximum	Unit	Notes
$V_{\rm CC}$	V <sub>CC</sub> relative to V <sub>SS</sub>	-0.3	+7.0	V	
$V_{\mathrm{T}}$	$\overline{ m DC}$ voltage applied on any pin excluding $V_{CC}$ relative to $V_{SS}$	-0.3	+7.0	V	
$T_{\mathrm{OPR}}$	Operating ambient temperature	-20	+70	°C	Commercial
T <sub>STG</sub>	Storage temperature	-55	+125	°C	
$T_{\mathrm{SOLDER}}$	Soldering temperature	-	+260	°C	10 sec max.
$T_{ m BIAS}$	Temperature under bias	-40	+85	°C	

Note:

Permanent device damage may occur if **Absolute Maximum Ratings** are exceeded. Functional operation should be limited to the Recommended DC Operating Conditions detailed in this data sheet. Exposure to conditions beyond the operational limits for extended periods of time may affect device reliability.

# **DC Thresholds** $(T_A = T_{OPR}; V_{CC} \pm 10\%)$

Symbol	Parameter	Rating	Tolerance	Unit	Notes
V <sub>SNSHI</sub>	High threshold at SNS resulting in MOD = Low	$0.05*\mathrm{V_{CC}}$	±0.025	V	
$V_{\mathrm{SNSLO}}$	Low threshold at SNS resulting in MOD = High	$0.04*\mathrm{V}_{\mathrm{CC}}$	±0.010	V	
$V_{\rm LTF}$	Low-temperature fault	$0.4*\mathrm{V_{CC}}$	±0.030	V	$V_{TEMP} \ge V_{LTF}$ inhibits/terminates charge
$V_{\mathrm{HTF}}$	High-temperature fault	$(1/3 * V_{LTF}) + (2/3 * V_{TCO})$	±0.030	V	$ V_{TEMP} \leq V_{HTF} \ inhibits \\ charge $
$V_{\rm EDV}$	End-of-discharge voltage	$0.4*\mathrm{V_{CC}}$	±0.030	V	$\begin{aligned} V_{CELL} < V_{EDV} & inhibits \\ fast & charge \end{aligned}$
$V_{ m MCV}$	Maximum cell voltage	$0.8*\mathrm{V_{CC}}$	±0.030	V	$V_{\rm CELL} > V_{\rm MCV}$ inhibits/ terminates charge
V <sub>THERM</sub>	TS input change for $\Delta T/\Delta t$ detection	-16	±4	mV	$V_{\rm CC}$ = 5V, $T_{\rm A}$ = 25°C
-ΔV	BAT input change for $-\Delta V$ detection	-12	±4	mV	$V_{\rm CC}$ = 5V, $T_{\rm A}$ = 25°C
PVD	BAT input change for PVD detection	-6	±2	mV	$V_{\rm CC}$ = 5V, $T_{\rm A}$ = 25°C

# **Recommended DC Operating Conditions (TA = TOPR)**

Symbol	Condition	Minimum	Typical	Maximum	Unit	Notes
$V_{\rm CC}$	Supply voltage	4.5	5.0	5.5	V	
$V_{BAT}$	Battery input	0	-	$V_{CC}$	V	
$V_{CELL}$	BAT voltage potential	0	-	$V_{\rm CC}$	V	V <sub>BAT</sub> - V <sub>SNS</sub>
V <sub>TS</sub>	Thermistor input	0	-	$V_{\rm CC}$	V	
$V_{TEMP}$	TS voltage potential	0	-	$V_{\rm CC}$	V	V <sub>TS</sub> - V <sub>SNS</sub>
V <sub>TCO</sub>	Temperature cutoff	0.2 * V <sub>CC</sub>	-	0.4 * V <sub>CC</sub>	V	Valid ΔT/Δt range
***	Logic input high	2.0	-	-	V	DCMD, INH
$V_{\mathrm{IH}}$	Logic input high	V <sub>CC</sub> - 0.3	-	-	V	TM <sub>1</sub> , TM <sub>2</sub> , DSEL, VSEL
	Logic input low	-	-	0.8	V	$\overline{\mathrm{DCMD}}, \overline{\mathrm{INH}}$
$ m V_{IL}$	Logic input low	-	-	0.3	V	TM <sub>1</sub> , TM <sub>2</sub> , DSEL, VSEL
V <sub>OH</sub>	Logic output high	V <sub>CC</sub> - 0.8	-	-	V	DIS, MOD, LED <sub>1</sub> , LED <sub>2</sub> , $I_{OH} \le -10 mA$
$V_{\mathrm{OL}}$	Logic output low	-	-	0.8	V	$\begin{array}{c} DIS,MOD,LED_1,LED_2,\\ I_{OL}\leq 10mA \end{array}$
$I_{CC}$	Supply current	-	1	3	mA	Outputs unloaded
$I_{SB}$	Standby current	-	-	1	μA	$\overline{INH} = V_{IL}$
I <sub>OH</sub>	DIS, LED <sub>1</sub> , LED <sub>2</sub> , MOD source	-10	-	-	mA	$@V_{OH} = V_{CC} - 0.8V$
$I_{OL}$	DIS, LED <sub>1</sub> , LED <sub>2</sub> , MOD sink	10	-	-	mA	$@V_{OL} = V_{SS} + 0.8V$
т	Input leakage	-	-	±1	μA	$\overline{\text{INH}}$ , BAT, V = V <sub>SS</sub> to V <sub>CC</sub>
$I_{\rm L}$	Input leakage	50	-	400	μA	$\overline{\mathrm{DCMD}}$ , $V = V_{\mathrm{SS}}$ to $V_{\mathrm{CC}}$
I <sub>IL</sub>	Logic input low source	-	-	70	μΑ	$\begin{aligned} TM_1, TM_2, DSEL, VSEL, \\ V = V_{SS} \text{ to } V_{SS} + 0.3V \end{aligned}$
$I_{IH}$	Logic input high source	-70	-	-	μΑ	$\begin{aligned} TM_1, TM_2, DSEL, VSEL, \\ V = V_{CC} - 0.3V \text{ to } V_{CC} \end{aligned}$
$I_{\rm IZ}$	Tri-state	-2	-	2	μA	TM <sub>1</sub> , TM <sub>2</sub> , DSEL, and VSEL should be left disconnected (floating) for Z logic input state

# Impedance

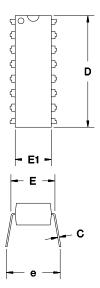
Symbol	Parameter	Minimum	Typical	Maximum	Unit
R <sub>BAT</sub>	Battery input impedance	50	-	-	ΜΩ
R <sub>TS</sub>	TS input impedance	50	-	-	ΜΩ
R <sub>TCO</sub>	TCO input impedance	50	-	-	ΜΩ
R <sub>SNS</sub>	SNS input impedance	50	-	-	ΜΩ

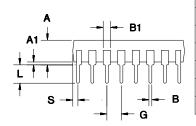
# Timing (TA = 0 to +70°C; VCC $\pm$ 10%)

Symbol	Parameter	Minimum	Typical	Maximum	Unit	Notes
$t_{\mathrm{PW}}$	Pulse width for DCMD and INH pulse command	1	-	-	μs	Pulse start for charge or discharge before charge
d <sub>FCV</sub>	Time base variation	-16	-	16	%	V <sub>CC</sub> = 4.75V to 5.25V
$f_{ m REG}$	MOD output regulation frequency	-	-	300	kHz	
$t_{ m MCV}$	Maximum voltage termination time limit	1	-	2	s	Time limit to distinguish battery removed from charge complete.

Note: Typical is at  $T_A = 25$ °C,  $V_{CC} = 5.0$ V.

# 16-Pin DIP Narrow (PN)

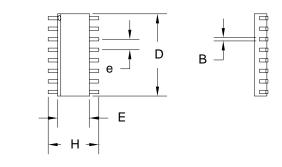


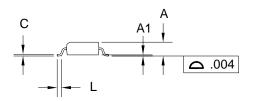


16-Pin PN (0.300" DIP)

	Inches		Millin	neters				
Dimension	Min.	Max.	Min.	Max.				
A	0.160	0.180	4.06	4.57				
A1	0.015	0.040	0.38	1.02				
В	0.015	0.022	0.38	0.56				
B1	0.055	0.065	1.40	1.65				
C	0.008	0.013	0.20	0.33				
D	0.740	0.770	18.80	19.56				
E	0.300	0.325	7.62	8.26				
E1	0.230	0.280	5.84	7.11				
e	0.300	0.370	7.62	9.40				
G	0.090	0.110	2.29	2.79				
L	0.115	0.150	2.92	3.81				
S	0.020	0.040	0.51	1.02				

# 16-Pin SOIC Narrow (SN)





16-Pin SN (0.150" SOIC)

	Inches		Millin	neters
Dimension	Min.	Max.	Min.	Max.
A	0.060	0.070	1.52	1.78
A1	0.004	0.010	0.10	0.25
В	0.013	0.020	0.33	0.51
C	0.007	0.010	0.18	0.25
D	0.385	0.400	9.78	10.16
E	0.150	0.160	3.81	4.06
e	0.045	0.055	1.14	1.40
Н	0.225	0.245	5.72	6.22
L	0.015	0.035	0.38	0.89

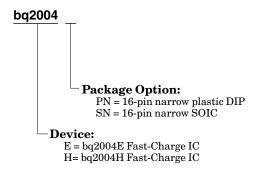
# **Data Sheet Revision History**

Change No.	Page No.	Description	Nature of Change
1	All	Combined bq2004E and bq2004H, revised and expanded format of this data sheet	Clarification
2	7	Separated bq2004E and bq2004H in Table 2, LED Output Summary	Clarification
3	5	Description of charge-pending state	Clarification
4			

Note:

Change 1 = Oct. 1997 B changes from Sept. 1996 (bq2004E), Feb. 1997 (bq2004H). Change 2 = Feb. 1998 C changes from Oct. 1997 B. Change 3 = Dec. 1998 D changes from Feb. 1998 C. Change 4 = June 1999 E changes from Dec. 1998 D.

# **Ordering Information**



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